Towards a Method to Determine the Glottal Formant Parameters of Voiced Speech without Time-Domain Reference

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Abstract — This paper presents an approach to estimate the glottal formant parameters of the voicing source in the frequency-domain. The method is based on a simplified pole-zero interpretation of the prevalent Liljencrants-Fant (LF) model of glottal flow, and gives approximations for a broad range of pulses shapes. An advantage of the method is that, unlike other methods, it does not rely on time-domain references.

Keywords — glottal source parameterisation, LF model, glottal formant

I INTRODUCTION

Parameterising the voicing source of speech has many applications in speech including: speaker identification [1], the synthesis of natural speech [2] and timbral modifications of voice quality [3].

One frequency-domain aspect of the voicing source is a region of increased spectral energy referred to as the glottal formant. A prevalent method for estimating the glottal formant parameters is to fit a model to the inverse filtered speech signal in the time-domain [4]. Following the convolution of the speech pulse with the vocal tract inverse filter, the various landmarks representing important time points of the glottal source are directly estimated from the time-domain waveform. These initial approximations are then refined by using an optimisation algorithm to minimise a suitable error criterion, e.g. time-domain least squares.

A second method to determine the glottal formant parameters exploits the phase properties of the glottal source to separate the maximum-phase glottal formant from the minimum-phase characteristics of the vocal tract resonances and glottal pulse return phase [5]. By appropriately positioning the analysis frame over the instant of glottal closure and calculating the zeros of the Z-transform, the open phase of the voicing source can be reconstructed from a subset of the calculated roots. The glottal formant parameters can then be determined by an analysis of this signal.

A pre-requisite of both of these methods is knowledge of the evolution of the glottal waveform in the time-domain. These instants can be obtained by the analysis of an electroglottograph signal recorded simultaneously with the speech signal [6], or via various algorithms to detect the epochs directly from the speech waveform, e.g. [7]. However, in the case of signals that have been phase corrupted, by e.g. non-linear phase recording equipment, these instants may not be recoverable without a phase-correction filter, the design of such a filter is not always feasible [8].

This paper develops a strategy for determining the glottal formant characteristics from a glottal pulse signal without reference to its time evolution. The outline of this paper is as follows: the following section gives the necessary background on the acoustic theory of speech production and a detailed description of a prevalent model of a pulse of the derivative glottal flow, the LF model. Here the glottal formant parameters are described along with its time-domain correlates. The third section shows how the LF model can be expressed as a high-order pole-zero filter, and develops an
approximation. Experiments validating this approach are given in the fourth section. The fifth section discusses the results yielded by these experiments. Conclusions are drawn in the final section, which also outlines the direction of future research.

II BACKGROUND

a) Acoustic Theory of Speech Production

The acoustic theory of speech production [9] views speech as the convolution of glottal flow signal with a vocal tract filter which is then radiated at the lips. In the Z-domain, the process can be represented as follows:

\[ S(z) = G(z)V(z)L(z) \]  

(1)

where \( S(z) \) represents the speech waveform, \( G(z) \) the glottal flow, \( V(z) \) the vocal tract filter, and \( L(z) \) represents lip radiation.

As lip radiation \( L(z) \) is usually modeled as a differentiating filter and the relationship between the speech chain components assumed linear, it is often combined with the glottal flow \( G(z) \) to form the derivative glottal flow \( G'(z) \).

b) The LF Model of the Glottal Flow Derivative

The LF model [10] represents the general flow shape of the glottal flow derivative over one glottal cycle and whose shape can be uniquely described with four parameters. The mathematical formula describing the LF model is a piece-wise function, consisting of two segments, the evolution of which can be seen in Fig. 1. The first segment is an exponentially increasing sine function, characterizing the glottal flow derivative from the instant of glottal opening \( t_o \), through the time axis at \( t_p \), to the instant of maximum negative extreme at \( t_e \).

At this point the second segment of the LF model, often referred to as the return phase, begins. This portion models the glottal closure as a modified exponential function which returns to zero at a rate determined by the steepness of the slope of the tangent to the function at \( t_e \). The distance of this tangent’s time axis intercept from \( t_e \) is called \( T_a \), and is referred to as the effective duration of the return phase. The total number of samples in the pulse is the pitch period, \( T_0 \). The value is related to the fundamental frequency \( f_0 \) and the sampling frequency \( f_s \) by the expression:

\[ T_0 = \frac{f_s}{f_0} \]  

(2)

In order to correctly place the pulse in time, the timing instants are calculated to be relative to the instant of glottal opening, i.e. \( T_0 = 0 \), \( T_p = t_p - t_o \), \( T_e = t_e - t_o \) and \( T_a = t_e - t_p \). Below are the mathematical equations describing time-domain LF model shape using these parameters:

\[ u_{LF}(n) = \begin{cases} E_0 e^{\alpha m} \sin \omega_0 n & 0 \leq n < T_e \\ \frac{E_e}{e^{T_a}} (e^{-\alpha(n-T_e)} - e^{-\alpha(T_e-T_c)}) & T_e \leq n \leq T_c \\ 0 & T_c \leq n < T_0 \end{cases} \]  

(3)

Among the different parameter sets that can be used to generate an LF model are the shape parameters. These parameters are the \( O_q \) the open quotient of the pulse, \( \alpha_m \) its asymmetry coefficient, and \( Q_a \) its return phase coefficient which can be expressed by the following equations:

\[ O_q = \frac{T_c}{T_0} \quad \alpha_m = \frac{T_p}{T_e} \quad Q_a = \frac{T_a}{(1 - O_q)T_0} \]  

(4)

These quantities are convenient because they bear a more meaningful relationship with the glottal behaviour than the timing parameters individually. For example, the open quotient is related to the time during which the glottal cycle is in its open phase during the pulse, and therefore an indication of the breathiness of the voice [11]. Another advantage is the normalised limits of these ratios.

c) Frequency-domain Features of the LF Model

In their work on the frequency spectra of the glottal models, Doval and d’Alessandro [12] describe the main frequency-domain features of the glottal source models. The salient spectral features of these models are the previously mentioned glottal formant and the spectral tilt. These features are referred to in Fig. 2.

The glottal formant is a band of increased spectral energy in the region of the voiced speech fundamental frequency. The term glottal formant is a misnomer as there is no resonance effect in the same manner as is with the vocal tract: rather, the frequency and degree of the spectral energy
This can be fully modelled as the impulse response of an unstable 2\textsuperscript{nd} order IIR filter, truncated at the point \( T_c \). The Z-transform of the open phase then takes the form:

\[
H_{op}(z) = \frac{1}{1 - 2e^{\alpha} \cos \omega_g z^{-1} + e^{2\alpha} z^{-2}} \quad (5)
\]

Recall that the parameters of this filter \( \omega_g \) and \( \alpha \) are the centre frequency and bandwidth of the previously mentioned glottal formant.

In order to completely represent the LF model as a pole-zero filter, it is desirable to construct a filter which will cancel the exponentially expanding sine wave beyond the point \( T_c \), to create the glottal closed phase. In the case of glottal waveforms that close abruptly, i.e. immediately after the instant of glottal closure, it can be shown that this cancelling filter can be constructed according to the following transfer function:

\[
H_{cp}(z) = 1 + b_{T_c+1} z^{-(T_c+1)} + b_{T_c+2} z^{-(T_c+2)} \quad (6)
\]

where

\[
b_{T_c+1} = e^{\alpha T_c} \frac{\sin \omega_g (T_c - 1) - 2 \cos \omega_g \sin \omega_g T_c}{\sin \omega_g} \quad (7)
\]

\[
b_{T_c+2} = e^{\alpha (T_c+1)} \frac{\sin \omega_g T_c}{\sin \omega_g} \quad (8)
\]

The FIR filter \( H_{cp} \) can be seen as a truncating/cancelling filter, whose numerous zeros impart a certain degree of spectral tilt and ripple onto the spectrum of the LF model.

In order to extend this representation to the more general case where the LF model pulse does exhibit a return phase, rather than cancel the expanding sine wave of the open phase, \( H_{cp} \) must instead supply it.

The LF model return phase is usually approximated by a low-pass IIR filter of the form [13]:

\[
H_{ret}(z) = \frac{1}{1 - \mu_{ret} z^{-1}} \quad (9)
\]

The impulse response of the filter \( H_{ret}(z) \) can be approximated by an FIR filter of sufficiently high order. Such a behaviour can be incorporated into the filter \( H_{cp} \) by augmenting it such that the \( n \)th term beyond \( b_{T_c+2} \) can be determined from the following equation:

\[
b_{T_c+2+n} = \mu_{ret}^n b_{T_c+2} \quad (10)
\]

Thus, the generalised \( H_{cp} \) filter which cancels the open phase and adds a return phase can be expressed:

\[
H_{cp} = 1 + b_{T_c+1} z^{-(T_c+1)} + b_{T_c+2} R(z) \quad (11)
\]
where
\[ R(z) = z^{-\tau} + p_{ret}z^{-(r+1)} + \ldots + p_{ret}^{n}z^{-(r+n)} + \ldots \]  
and \( \tau = T_e + 2 \).

A complete pole-zero model of an LF model pulse can therefore be expressed:
\[ G_{LF}^t(z) = \frac{H_{cp}(z)H_{op}(z)}{1 + b_{T_e+1}z^{-(T_e+1)} + b_{T_e+2}R(z)} \]  
\[ = 1 + \frac{e^{\alpha}}{2 \cos \omega g z^{-1} + e^{2\alpha}z^{-2}} \]  
(13)

From the above derivation, it follows that the open phase filter parameters \( H_{op} \) can be estimated given an approximation of the filter \( H_{cp} \). However, due to the complexity of the filter given by (13), in order to produce a tractable approximation, the approximation developed here is based on the simpler transfer function of the abruptly closing LF model, whose truncating filter is given by (6).

A filter which approximates the behaviour of the impulse response of \( H_{cp} \) near the point of glottal closure is a filter of the form:
\[ \hat{H}_{cp}(z) = 1 + \beta z^{-1} \]  
(14)
where \( \beta = \frac{b_{T_e+1}}{b_{T_e+1}} \). This ratio can be expressed as:
\[ \beta = e^{\alpha} \frac{\sin \omega g (T_e - 1)}{\sin \omega g k_e} - 2 \cos \omega g \]  
(15)

The range of values of \( \beta \) are dependent upon the variables \( \omega g, \alpha \) and \( T_e \). Surveying the range of \( \beta \) values given the physically reasonable limits of these parameters indicates that a good estimate for \( \beta \) is \(-1.01\). The similarity between a glottal pulse transfer function given by \( G_{lf}(z) = \frac{H_{pf}(z)}{H_{op}(z)} \) and the approximation \( \hat{G}_{lf}(z) = 1 + 10^{-1}z^{-1} \) can be seen in Fig. 3.

From this approximation, it follows that the coefficients describing the glottal formant open phase filter can be estimated by a two-pole autoregressive analysis on the glottal source that has been pre-emphasised by the inverse of the filter given by (14). However, as \( \beta \) will be less than \(-1\), this filter would constitute an unstable filter. In order to avoid this issue, the stable equivalent of this filter, i.e. where its pole is reflected about the unit circle, can be as effective once the signal is time reversed. For the analysis algorithm, the signal frames are pre-emphasised by the following filter:
\[ H_{pe}(z) = \frac{1}{1 + \frac{1}{\beta} z^{-1}} \]  
(16)

IV Experiment

An experiment was carried out in order to test the efficacy of the above approximation to determine the glottal formant parameters. At a sampling frequency of 10kHz, a broad range of different LF model pulses were generated, the parameters of which are listed in Table 1. In order to minimise the effect of the spectral tilt on the estimation of the glottal formant, each analysis pulse was first decimated to 2000 Hz.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Range</th>
</tr>
</thead>
<tbody>
<tr>
<td>( f_0 )</td>
<td>80 : 20 : 200 (Hz)</td>
</tr>
<tr>
<td>( Q_g )</td>
<td>0.3 : 0.05 : 0.9</td>
</tr>
<tr>
<td>( \alpha_m )</td>
<td>0.67 : 0.05 : 0.9</td>
</tr>
<tr>
<td>( Q_a )</td>
<td>0 : 0.05 : 1</td>
</tr>
</tbody>
</table>

Table 1: All LF model parameter configurations used for synthetic testing.

In order to cancel the effects of \( H_{cp} \), the frame was then time-reversed and pre-emphasised using the filter \( H_{pe} \), with \( \beta \) chosen to be \(-1.01\). A Hann window was then applied to the pulse, which then underwent second-order discrete all-pole (DAP) analysis [14], yielding an IIR filter.

The roots of the filter polynomial were calculated in order to locate the poles on the Z-plane. The coordinates of the poles underwent a Cartesian to polar conversion, yielding the estimates of the glottal formant center frequency \( \hat{\omega}_g \) and the exponential of its bandwidth \( \hat{\alpha} \). As DAP analysis will generally find stable poles, i.e. within the unit circle, the value of \( \hat{\alpha} \) is multiplied by \(-1\) in order to determine the pole’s maximum-phase reflection.

The errors of \( \hat{\omega}_g \) and \( \hat{\alpha} \) are compared to their original values by the following formulae:
\[ E_{\omega g} = \frac{\hat{\omega}_g - \omega g}{\omega g} \]  
\[ E_{\alpha} = \frac{\hat{\alpha} - \alpha}{\alpha} \]  
(17)

The error associated with \( \hat{\omega}_g \) is calculated as the percentage error of the pole radius, rather than the percentage error of \( \hat{\alpha} \). This is because the small values of \( \alpha \) tends to exaggerate even slight deviations from its true value.
V Discussion of Results

The results of the experiment are given in Figs. 4 and 5. Each figure plots the percentage error associated with $\alpha$ and $\omega_g$ against the number of waveforms analysed exhibiting that error.

As can be seen in Fig. 4, the $E_{\alpha}$ is generally slightly underestimated. The accentuation of the low frequencies of the filter $H_{pe}$ tends to “dull” spectral peaks of the glottal formant, reducing the bandwidth. This slight affect can be seen in Fig. 6, where increased return phase coefficients, which are known to exhibit broader glottal formants (see [12]), are shown to be correlate with slightly overestimated bandwidth values.

Contrary to $E_{\alpha}$, the values of $E_{\omega_g}$ exhibit a far wider degree of variation. The approximation tends to underestimate the glottal formant centre frequency.

The main source of error can most likely be attributed to the coarse approximation of the closed phase truncation filter $H_{pe}$. This filter necessarily boosts the low frequencies in order to compensate for the truncation of the open phase filter, but seems to be too crude in the cases where a more gradual return phase is required to be compensated. Indeed, even in cases of abrupt glottal closures, $\omega_g$ values are underestimated, see Fig. 7. Clearly, “one-size-fits-all” pre-emphasis is inadquate for the precise determination of the glottal formant parameters across the variety of voice source signals that may be encountered.

As can be seen in Fig. 7, more localised results are exhibited for glottal pulses that are characterised by a null return phase coefficient value. As the return phase coefficient grows larger, the results become more spread, consistently bias towards underestimating the true $\omega_g$ value. It is thought that this is due to the lack of a representation of the return phase within the approximation filter $H_{pe}$. Additionally, this filter is constructed from an approximation to the signal discontinuity required to truncate the open phase expanding sine wave at the point of glottal closure. The neglect of the initial filter coefficient of $H_{pe}$ would introduce some low frequency content which may also distort the subsequent auto-regressive analysis.

The goal of this research is to develop a method of determining the glottal formant without reference to the time-domain shape of the signal, yet the placement of the source signal within the frame affects the results in a subtle fashion. The experiment above was repeated for different pulse shapes placements, and the results obtained were broadly similar.

VI Conclusion

This paper presents a method to estimate the glottal formant parameters, without the reliance on the specific time information of the signal. This was achieved first by developing a transfer function whose poles and zeros represent the LF model. The denominator and numerator of the transfer
function represent separately the open and return/closed phase of a modified LF model respectively. In order to estimate the glottal formant parameters, characterised in the open phase of the LF model, a first order approximation to the numerator was developed, allowing an estimation of the glottal formant parameters.

The results of this method revealed that the estimations of the formant bandwidth are generally less than ±5%, while the error is larger for the glottal formant centre frequency. It is believed that this is mainly due to a lack of generality in the first order approximation. An adaptive strategy using other available signal information (e.g. the amplitude difference between the first two harmonics which has been shown to be able to estimate $O_q$ for moderate values [15]) may lead to better pre-emphasis filters.

Additionally, it is believed that this method to determine the glottal formant can be used to better inform an inverse filtering strategy regarding the contributions of the glottal source. A novel method of inverse filtering utilising such an approach is currently under development.

References


